

**Presented by:
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at**



Using Simulation to Estimate MOSFET Junction Temperature in a Circuit Application

Presented by:

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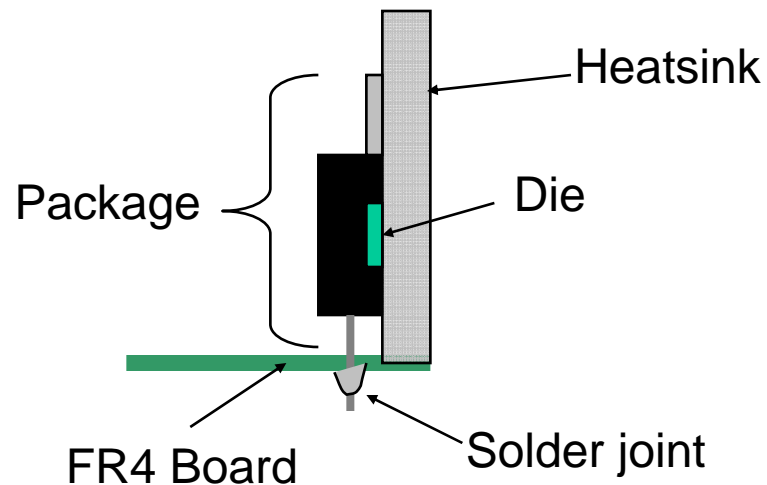
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Agenda

- Definition of Electro-Thermal Simulation
- Simulation Tools and Methods
- Methods of Estimating Die Temperature
- Creating Quasi-Dynamic MOSFET Model
- Model Generation
- Example Application
- Conclusion

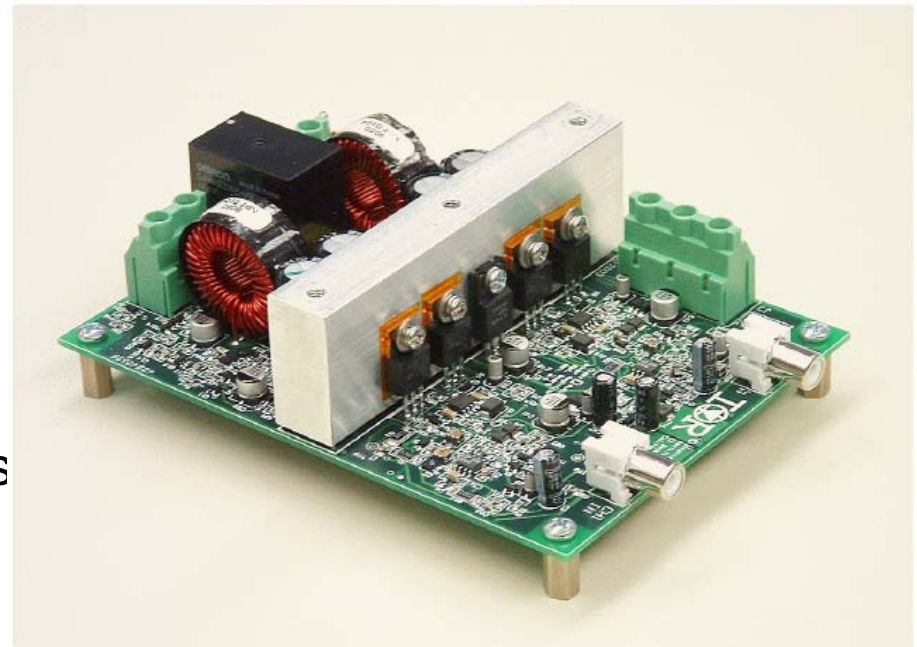
Electro-Thermal Simulation

- Purpose of Electro-Thermal Simulation is to predict MOSFET junction for a given application.



Electro-Thermal Simulation

- Applications
 - Solenoid drivers
 - Motor drive
 - Lighting ballast
 - DC/DC converters
 - Switch mode power supplies
 - Class D amplifiers



Simulation Tools and Methods

- Tools

- Simplorer (Ansoft) - Circuit/System simulator with VHDL-AMS hardware description language
- Saber (Synopsis) - Circuit/System simulator with VHDL-AMS and MAST hardware description languages
- Spector (Cadence) - Circuit/System simulator with Verilog-A hardware description language
- PSPICE (Cadence) – Defacto standard in circuit simulation.

Simulation Tools and Methods

Method

- ❑ Implementing model in the hardware description language
- ❑ Implementing model using equations and macro modeling

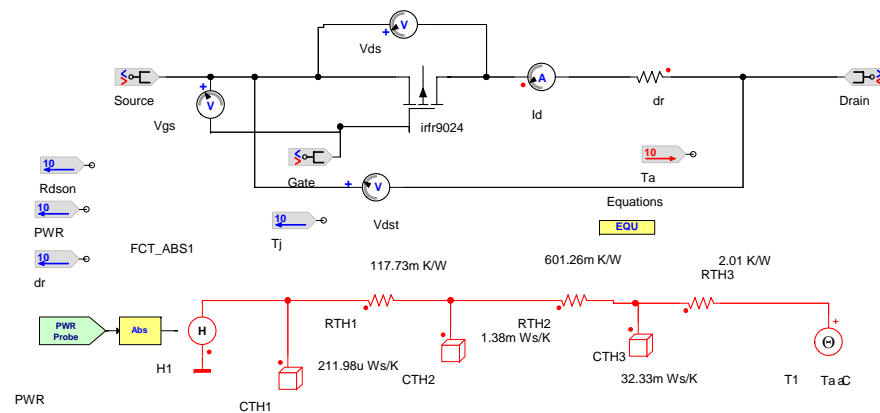
```
library ieee; use ieee.std_logic_1164.all;
use ieee.electrical_systems.all;
```

```
entity comparator is
  port ( terminal a : electrical;
        signal d : out std_ulogic );
end entity comparator;
```

```
architecture ideal of comparator is
  constant ref_voltage : real := 5.0;
  quantity vin across a;
begin
```

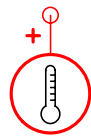
```
  comparator_behavior : process is
  begin
    if vin > ref_voltage / 2.0 then
      d <= '1' after 5 ns;
    else
      d <= '0' after 5 ns;
    end if;
    wait on vin'above(ref_voltage / 2.0);
  end process comparator_behavior;
```

```
end architecture ideal;
```



Simulation Tools and Methods

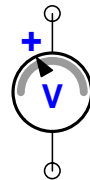
- Quasi-Dynamic MOSFET model implementation
 - ❑ Macro modeling with use of linking equations
 - ❑ Using a multi domain simulator that allows for Electro-Thermal simulation.
 - Volts and Amps
 - Heat flow (Watts) and temperature



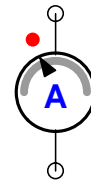
Thermometer



Heat



Voltmeter



Ammeter

Methods of Estimating Die Temperature

- Methods of estimating MOSFET die junction temperature
 - Equation based + Thermal Impedance curve

$$P = I * V * D$$

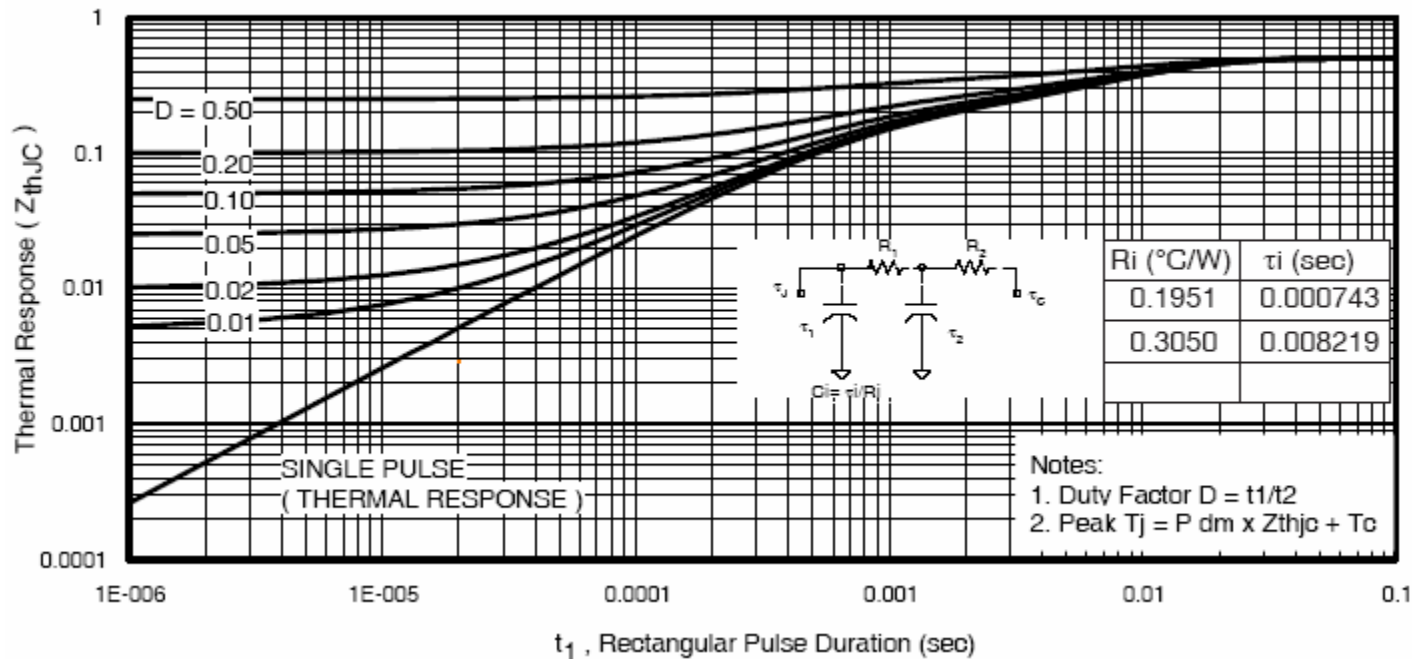
Where:

I = average current during the conduction cycle

V = equivalent voltage across the device during the conduction cycle

D = duty cycle

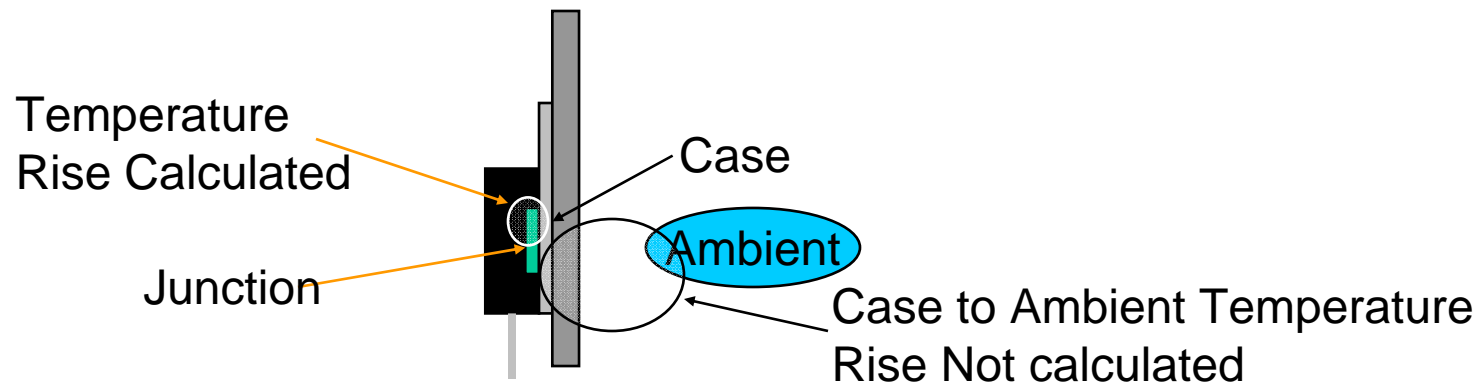
Methods of Estimating Die Temperature



- Use Power calculated with $P=I*V*D$
- Use pulse width and duty cycle to determine Z_{th} (thermal impedance) from device thermal impedance curve
- Temperature rise ($\Delta T_{junction}$)= $Z_{th} * P$

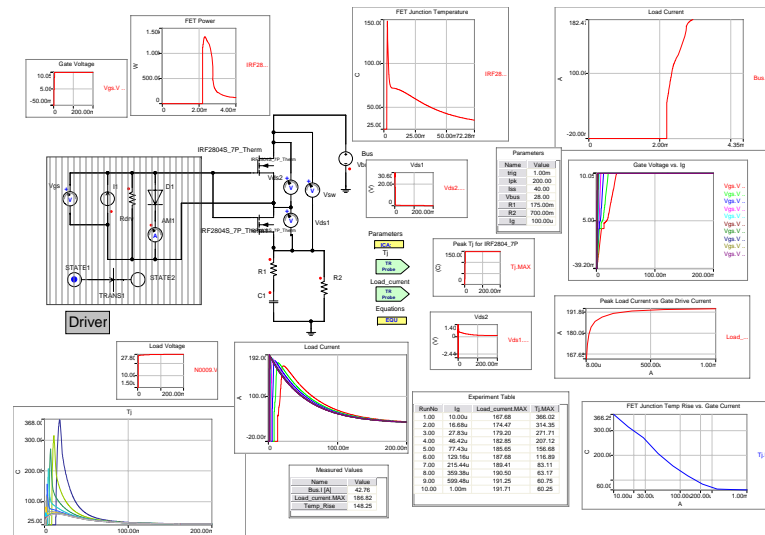
Methods of Estimating Die Temperature

- Limitations of the equation based junction temperature estimate
 - ❑ Only temperature rise from junction to case is taken into account. Neglects case to ambient temperature rise.
 - ❑ Assumes the power pulse is an ideal square edged pulse train.
 - ❑ It does not allow for transient thermal response.



Methods of Estimating Die Temperature

- **Simulator based MOSFET junction temperature estimate**
 - ❑ Uses circuit simulation to calculate junction temperature in an application
 - ❑ The circuit can be arbitrary
 - ❑ Transient thermal response is calculated
 - ❑ Component parameters change with temperature



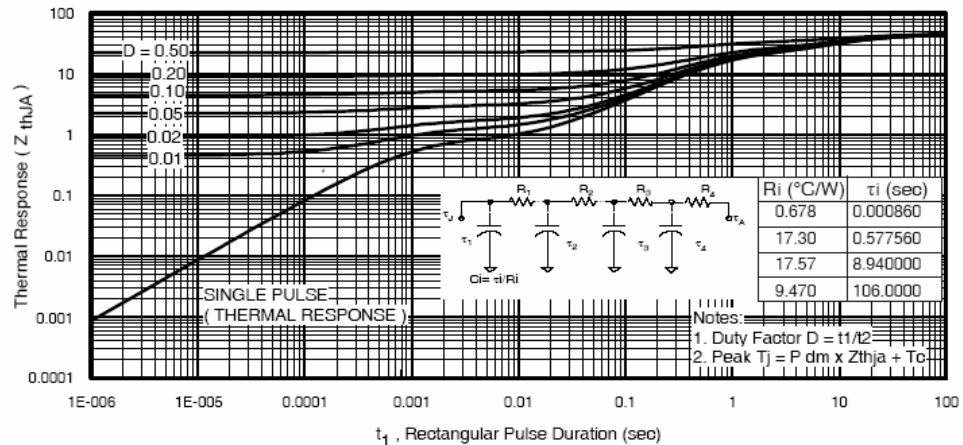
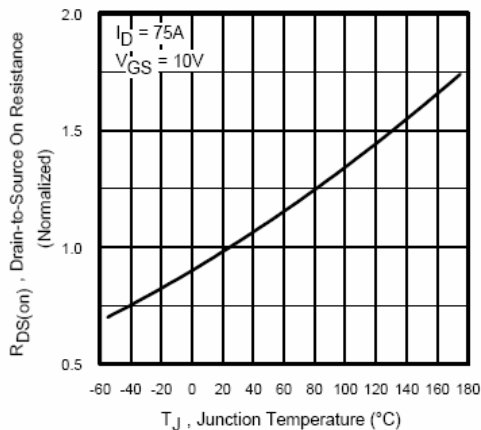
Methods of Estimating Die Temperature

- Assumptions made for junction temperature estimates using simulation
 - ❑ No other source of heat considered (Temperature rise due to self heating only)
 - ❑ Only MOSFET $R_{DS(on)}$ and threshold voltage changes with temperature
 - ❑ Since simulation solves Ordinary Differential the junction is assumed to be a point source of heat.

Creating Quasi-Dynamic Thermal MOSFET Model

- Gathering information:
 - ❑ 25C Spice Model of MOSFET
 - ❑ Datasheet information
 - $R_{DS(on)}$ vs. Temperature curve
 - Thermal Impedance Curve with thermal RC ladder network

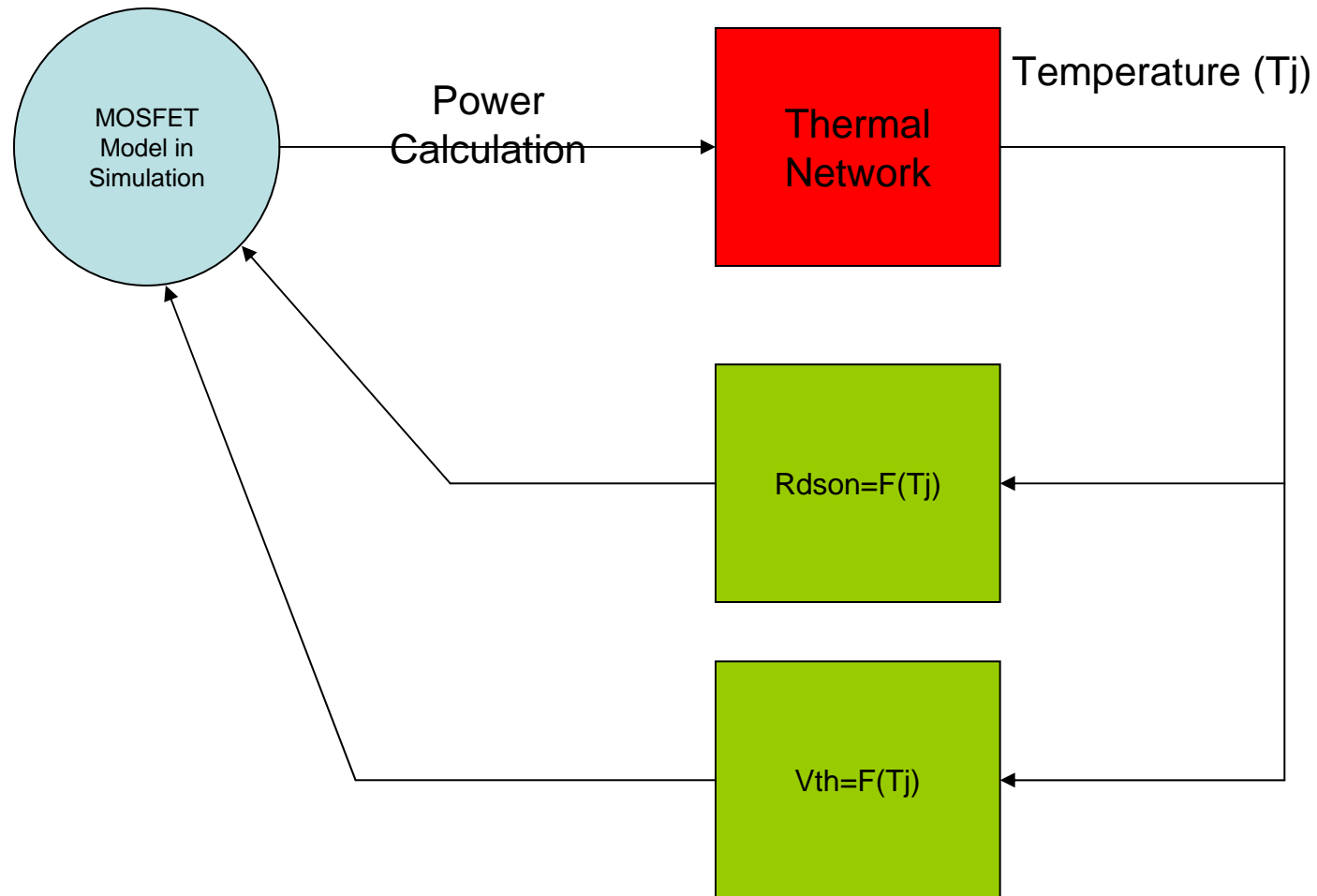
```
.SUBCKT irf1404 1 2 3 * SPICE3 MODEL WITH
THERMAL RC NETWORK
*****
* Model Generated
by MODPEX * *Copyright(c) Symmetry Design
Systems* * All Rights Reserved * * UNPUBLISHED
LICENSED SOFTWARE * * Contains Proprietary
Information * * Which is The Property of * *
SYMMETRY OR ITS LICENSORS * *Commercial Use
or Resale Restricted * * by Symmetry License
Agreement * *****
* Model
generated on April 2, 01 * MODEL FORMAT: SPICE3
* Symmetry POWER MOS Model (Version 1.0) *
External Node Designations * Node 1 -> Drain * Node
2 -> Gate * Node 3 -> Source M1 9 7 8 8 MM L=100u
W=100u .MODEL MM NMOS LEVEL=1 IS=1e-32
+VTO=3.74133 LAMBDA=0.00250986 KP=514.947
+CGSO=7.17952e-05 CGDO=1.60578e-08 RS 8 3
0.00282867 D1 3 1 MD .MODEL MD D IS=1.89845e-
10 RS=0.00218742 N=1.20398 BV=40 +IBV=0.00025
EG=1.2 XTI=1.85712 TT=2.00014e-05
+CJO=5.42237e-09 VJ=2.67939 M=0.566441 FC=0.1
6781
=0
```



Creating Quasi-Dynamic Thermal MOSFET Model

- 25C Spice Model
 - ❑ Characterized to the datasheet
 - ❑ Does not change performance characteristics as power is calculated
 - ❑ Used as base model for Quasi-Dynamic MOSFET model

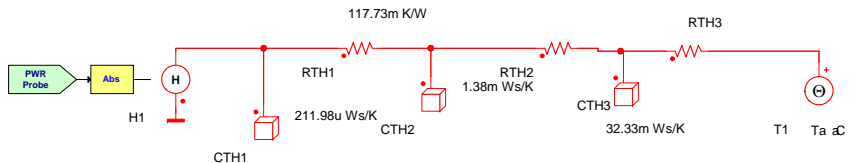
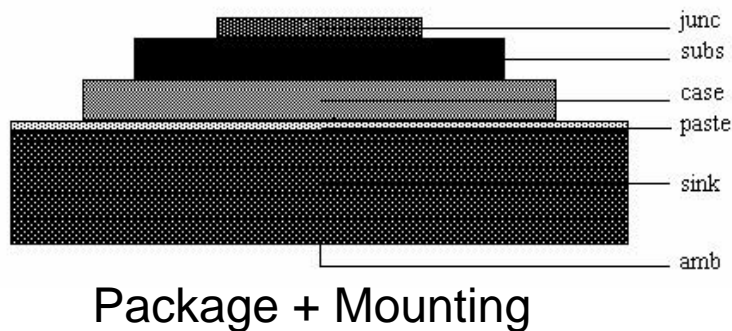
Creating Quasi-Dynamic Thermal MOSFET Model



Model Generation

▪ Ladder Network

- A thermal RC network used to model the dynamic thermal behavior of the package + mounting system.



Equivalent RC Ladder Network

The ladder network can be synthesized from the thermal impedance curve or is given by the MOSFET manufacturer

Model Generation

- Tying the thermal model to the 25C Spice model
 - ❑ Create the equation that represents $R_{DS(on)}$ vs. temperature

$$\frac{dR_{dson}(T_j)}{dT_j} = R_{dson}(25C) * (2 * a * T_j + b)$$

$$dR_{dson} = R_{dson}(25C) * (2 * a * T_j + b) * (T_j - 25)$$

← This expression gets implemented in the model

Note: a, b and c are calculated via a curve fitting routine.
The R_{dson} vs Temperature curve is assumed to be quadratic.

Model Generation

- Create the voltage source that represents the temperature dependence of V_{th} (threshold voltage)

$$V_{th}(T_j) = -0.007 * (T_j - 25) \longrightarrow \text{Expression used in model.}$$

The voltage source is in series with the MOSFET's gate.

Model Generation

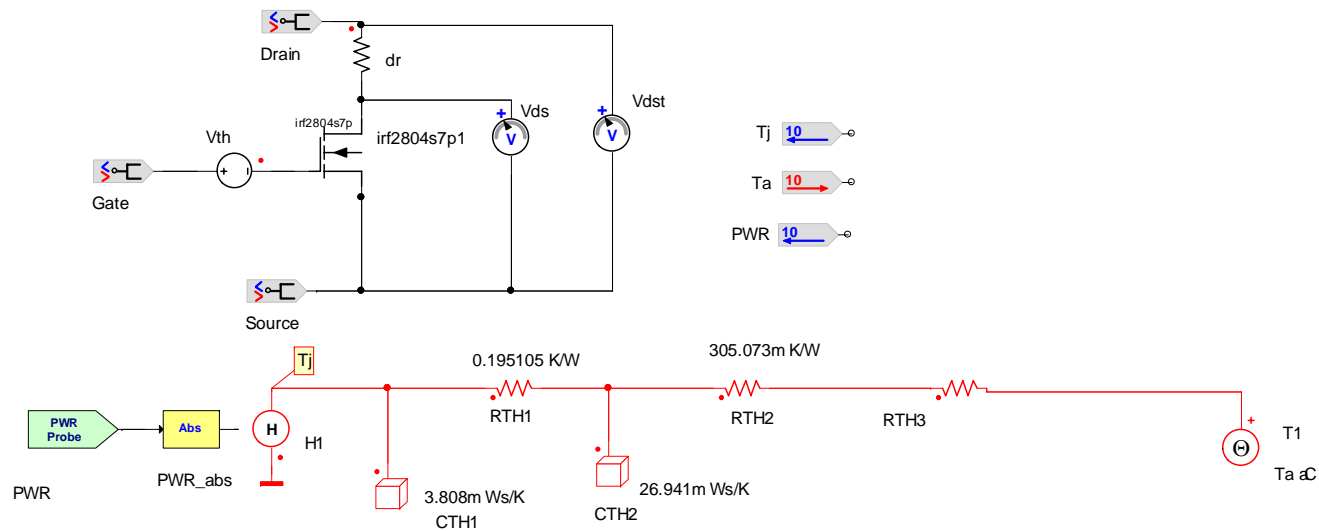
- Calculating the power in the MOSFET for use in the thermal network.

$$P = I_d * V_{ds}$$

This calculated power is the source for the thermal network.

Model Generation

■ Putting it together



Equations

EQU

$$dt := Tj.T - Ta$$

$$Tj := Tj.T - 273.15$$

$$\text{if}(Vds.V < 0.1) \{ Rdson25 := \text{abs}(Vds.V / dr.I) \} \text{ else } \{ dr := 1m \}$$

$$\text{if}(Vds.V < 0.1) \{ dr := (7.41u * Tj + 3.519m) * dt * Rdson25 \} \text{ else } \{ dr := 1u \}$$

$$PWR := PWR_abs.VAL$$

$$Vth := -7m * (Tj - 25)$$

Model Generation

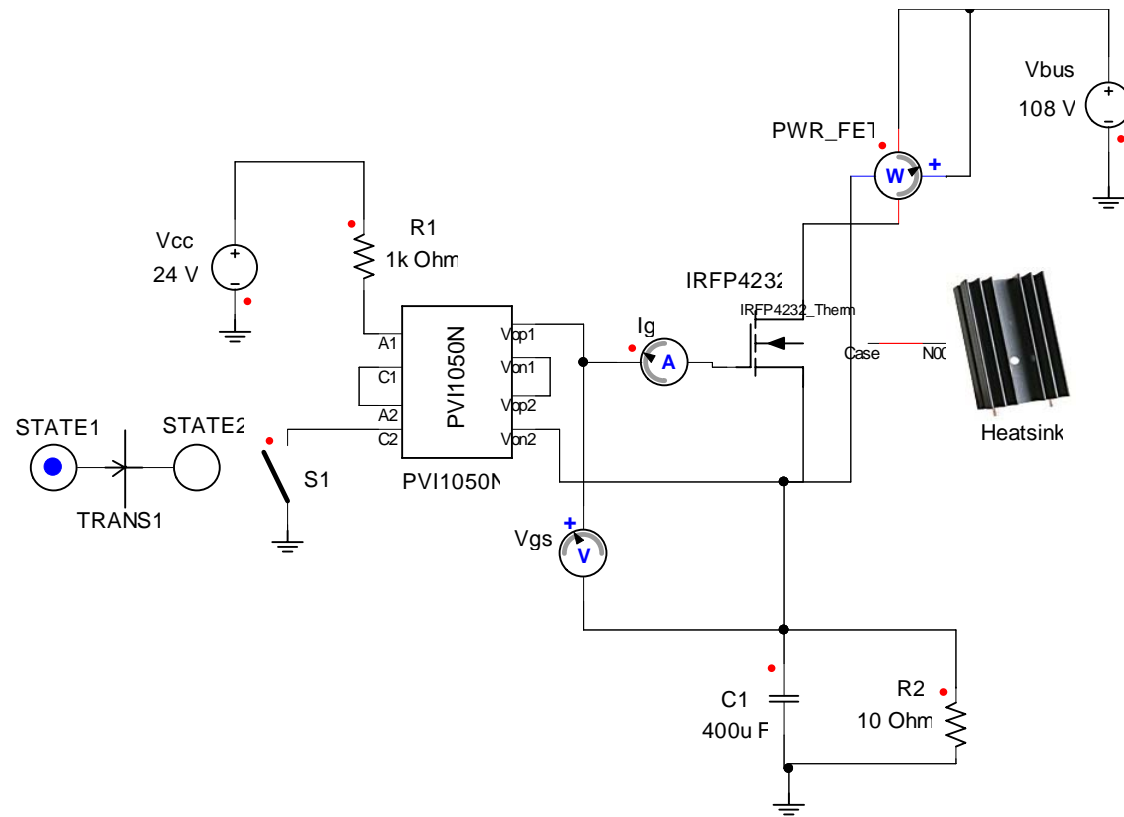
- Final model
 - ❑ 25C Spice model
 - ❑ Added voltage source Vth in gate implements $V_{th}(T_j)$
 - ❑ dr implements
 - ❑ Vds and the current in dr are used to calculate $R_{DS(on)}$ 25C
 - ❑ Vdst and the current in dr is used to calculate the total power
 - ❑ PWR_abs is used to insure that the thermal network is driven with positive power.

Example Application

- High side switch
- MOSFET being driven by a opto isolated drive
 - ❑ Very low drive current capability
- Load is capacitive
- Issue: How does driving this load effect the junction temperature of the MOSFET

Example Application

- Simulation Schematic



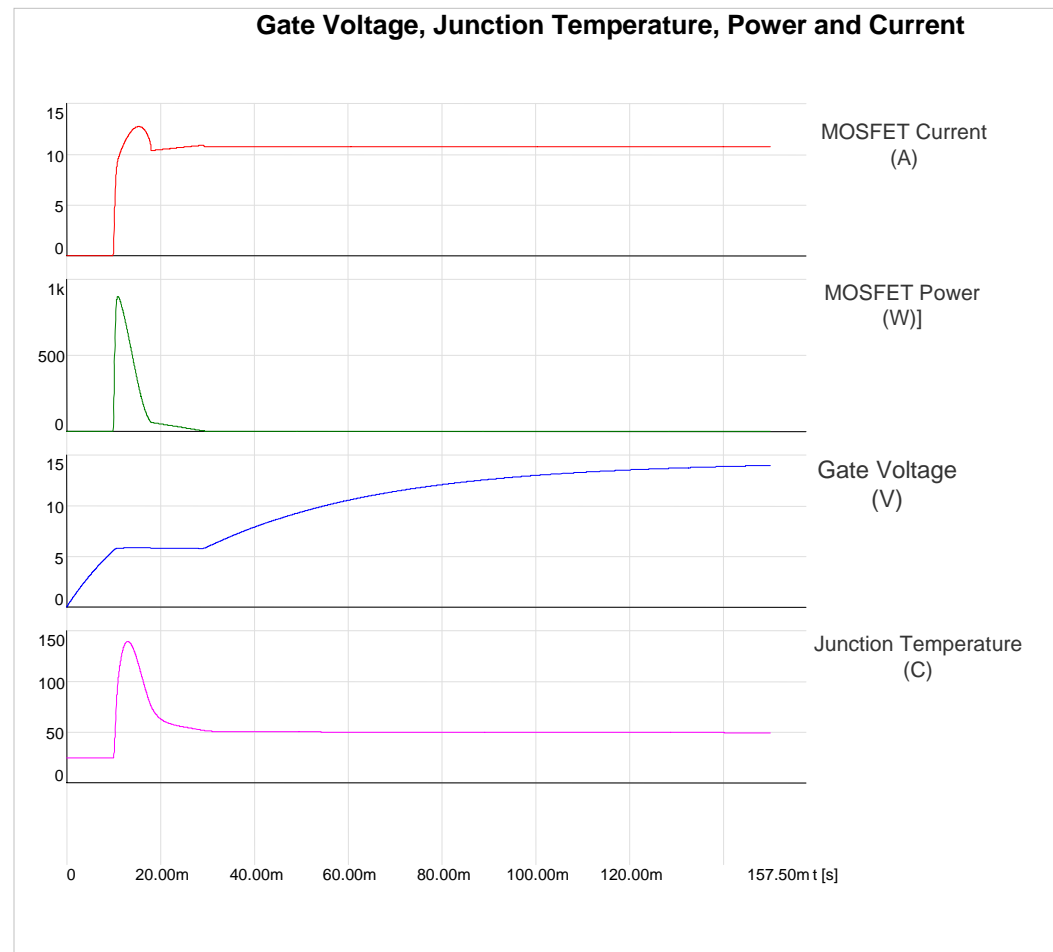
Example Application

- Assumptions

- Tambient=25C
- Heat sink is modeled as just a thermal resistor
- C1 & R2 represent a load system i.e. power supply
- Ig, Vgs, PWR_FET, States 1 & 2, Trans1 and S1 are measurements, input stimulus and ideal switch

Example Application

- Results



Conclusion

- Electro-Thermal simulation allows for analysis in both electrical and thermal domains
- Quasi-Dynamic Thermal MOSFET model allows for self-heating to alter $R_{DS(on)}$ and V_{th} during simulation as a function of temperature
- Quasi-Dynamic Thermal MOSFET Model generation is a data gathering task
- The example shows why it is difficult to switch a capacitive load with an opto-driver and a MOSFET due to the excessive junction temperature spike during turn-on.